MX506

Advance

November 1989



Single Chip LMR Audio Processor

Features

- AGC Amplifier
- Digital Gain Controls
- RX/TX Switching & Filtering
- Squelch Filter
- TX VOGAD
- Serial Bus for all Chip Functions

Benefits

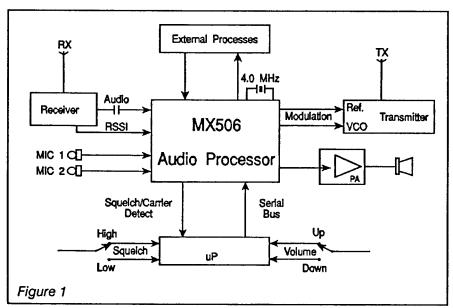
- Serves both Hand-heid and Mobiles
- Improves Range
- Saves Space, Power, Cost

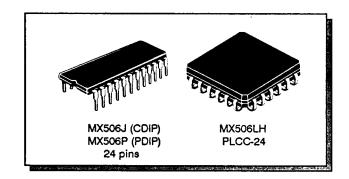
Description

The MX506 is a microprocessor-controlled, single-chip device containing ALL the circuit elements necessary to perform the audio functions of a mobile (or portable) radio system.

Each function in the signal path can be addressed or bypassed -- providing "real time" dynamic control -- through an externally produced serial control word.

This half-duplex device is comprised of two signal stages. The Pre-process path sets the incom-





ing audio (RX or TX) to levels and frequencies that can be used together with auxiliary systems such as Scrambling, Sub-Audible tones, or In-band data signaling.

This path can be output to external processes or internally routed to the Post-process path.

The Post-process path can adjust and prepare the input audio (either internal or external) for output to the chosen transmitter driver or loudspeaker amplifier.

> The MX506 can operate on Voice, Direct Digital or Tone data, and is compatible with FM, AM and SSB type transceivers. Digital gain elements are onchip for dynamic control and balance of signal levels during manufacturing test and operation.

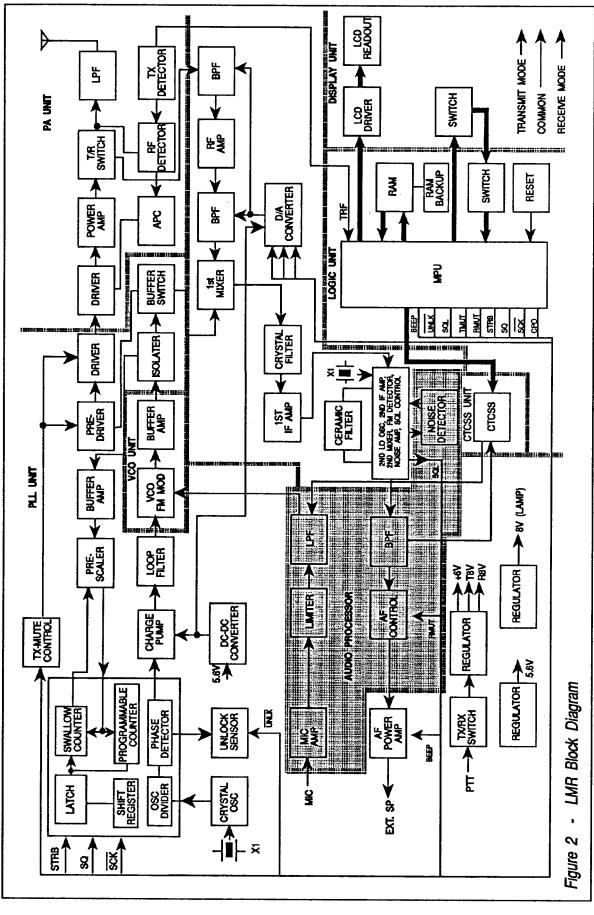
> System squelch is derived from either the input audio signal or the Received Signal Strength Indicator (RSSI) in the radio.

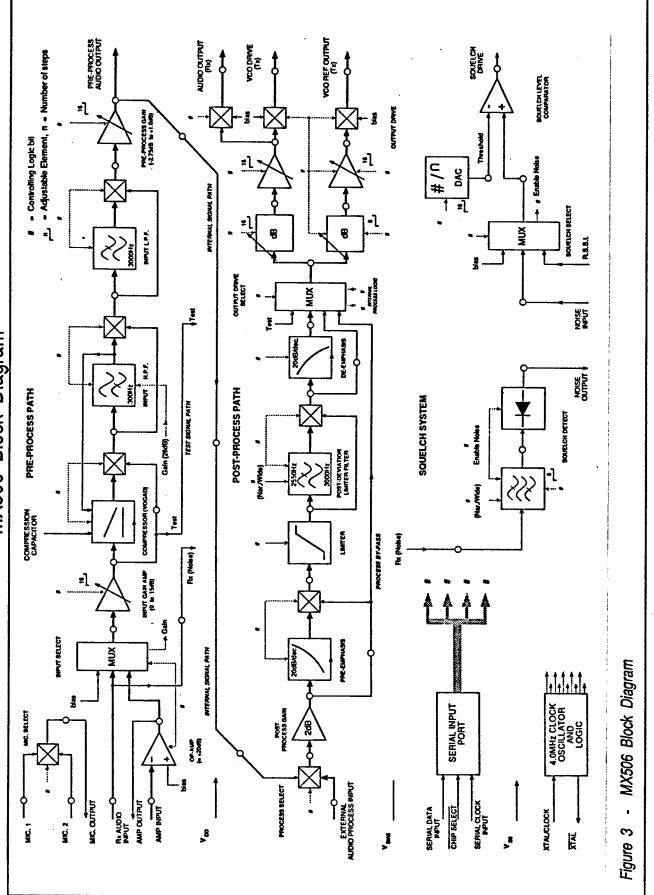
> This low-power 5 volt CMOS device is available in 24pin Cerdip, PDIP and PLCC.

> The PC5060, an evaluation printed circuit board, will be available to assist in MX506 application design.

Block Diagram for a Typical Land Mobile Radio







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MX506 Block Diagram

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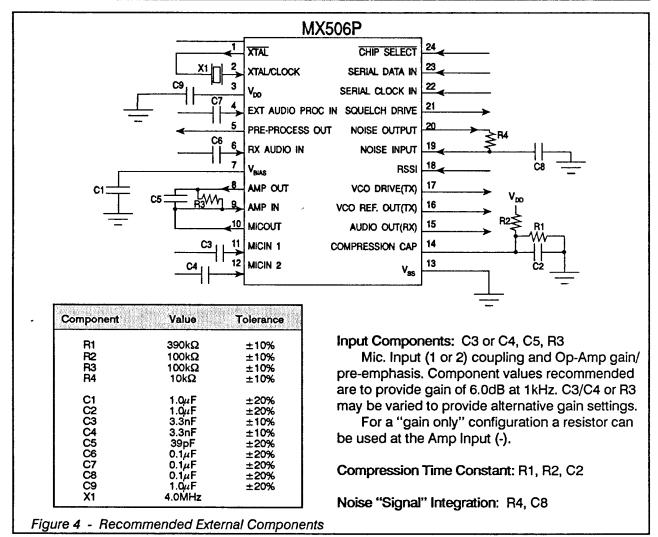
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Pin Function Chart

1. /

Pin	Function	
1	Xtal: The output of	the 4.0MHz on-chip clock oscillator.
2	Xtal: This is the inpur included on-chip. A See Figure 4.	t to the on-chip 4.0MHz clock oscillator inverter. All oscillator components are n external 4 MHz Xtal or externally derived clock should be connected here.
3	V _{DD} : Positive supply	y rail. A single, stable +5 volt supply is required.
4	operations. It is sele	ess Input: This is the analog input to the Post-Process Path from external audio cted by input serial data. When not selected, it is connected internally to the Output via the Internal Signal Path. Inputs to this pin should be a.c. coupled be Figures 3 and 4.
5	Pre-Process Audio output can be direct	Output : The analog output to external audio operations. The signal at this ted through the Internal Signal Path to the Post-Process Path. See Figure 3.
6	Coupling via capacit	e input from the radio receiver demodulator. This input, which requires a.c. or C6, is selected by serial data. Audio at this input will be available for use as ise source. See Figures 3 and 4.
7	V _{BIAS} : The output of decoupled to V _{SS} wi	the on-chip analog bias circuitry, internally set to $V_{DD}/2$. This pin should be ith capacitor C1. See Figure 4.
8 9	amp amp	n external components, this amplifier can be used as a microphone pre- blifier and pre-emphasis circuit. For a "gain only" configuration a serial stor may be employed at the Amp Input (-). See Figure 4.
10	Microphone Output: It should be connec	This is the output of the microphone multiplexer, selected by serial input data. ted to the Op-Amp Input (-). See Figures 3 and 4.
11 12	Mic. 1 Input: The Mic. 2 Input: inpu	se separate microphone audio inputs are individually selected by the serial It data. See Figure 3.
13	V _{ss} : Negative suppl	ly rail (GND).
14	Compression Capa compression time compression time compression time compression time complexity of the time complexity of time complexity of the time complexity of the time complexity of the time complexity of time com	citor: External components connected to this pirt provide the required onstant. See Figure 4.
15	selected. When in p	is is the received audio output from the Post-Process Path. This output is data owersave it is held at V _{BIAS} .
16	VCO Ref. Drive Outputis data selected. Wh	ult: This output is used to drive the modulation reference oscillator. This output en in powersave it is held at V _{BIAS} .
17	VCO Drive (TX) Outp powersaved it is hele	ut : This is the output drived to the Modulation VCO. It is data selected. When d at V _{BIAS} .
18	RSSI: This is the inpu Indicator. It is a data	t to the Squelch Selection Circuitry from the radio's Received Signal Strength -selected input.
19	Noise Input: The ne integrated by externa also be used.	oise level can be applied to this pin. The noise level is the Noise Output al components, as shown in Figure 4. An externally produced noise level can
20	d.c. level that can be	the output of the on-chip "squelch noise rectifier." It is a half-wave rectified applied to the Noise Input with external integrating components. This output by an external signal detector circuit. This output level is at V _{BIAS} for no input.
21	threshold level from t	-compatible output. The inputs to the comparator are the logically selected the Digital-to-Analog converter and the selected noise input. This output will he selected noise is greater than the set threshold, and a logic "0" when less

Pin	Function
	than the set threshold.
22	Serial Clock: This is the externally produced serial data loading clock input (See Figure 5). This input has an internal $1M\Omega$ pullup resistor.
23	Serial Data: This is the controlling 47-bit serial data input. When Chip Select is kept at logical "0," serial data is entered at this pin, bit 46 first and bit 0 last. Details concerning the allocation and function of serial data bits is on pages 6-9. Data load timing should be carried out as shown in Figure 4. This input has an internal 1 M Ω pullup resistor.
24	Chip Select: This controls data loading. During serial data loading this input should be operated as shown in Figure 5. New data is latched on the rising edge of this waveform. This input has an internal 1 M Ω pullup resistor.



Layout Recommendations

Audio Microcircuit performance will be affected by external noise.

All external components should be kept as close to the device as possible. Tracks to the device should be kept short, particularly to the Audio and VBIAS inputs. A "ground plane" connected to VSS will help to eliminate external pick-up.

Make sure that all inputs are free from noise.

Xtal/clock and digital tracks should be kept well away from analog circuitry. Analog inputs and outputs should be screened wherever possible with high-level outputs isolated from very low-level inputs.

Circuit Descriptions and Serial Control Information								
Control bits	Element	Notes						
<u>46</u> <u>45</u> 0 0 0 1 1 0 1 1	<u>Squelch Source Selector</u> - Internal Bias Path - Integrated external Noise input - RSSI Input	Enable Detector Filter and Rectifier Enable Detector Filter and Rectifier Enable Detector Filter and Rectifier Do not use this setting. Itfs is for future use.						
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	$\begin{array}{rrrrrrrrrrrrrrrrrrrrrrrrrrrrrrrrrrrr$	The fine squelch adjustment from the Digital-to Analog converter. These threshold levels are used as a comparison with the selected input noise voltage (bits 46 and 45). Variation in V_{pp} will produce variation in threshold levels.						
<u>40</u> 0 1	<u>Squelch Filter (Narrow/Wide)</u> - Narrow (fc = 18kHz ± 6.5kHz) - Wide (fc = 25kHz ± 8.5kHz)	For use in narrow or wide channel systems. The squelch function is set by bits 46 and 45.						
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	Squelch Filter (Gain) - Gain Set -3.0dB -2.0dB -1.0dB 0dB 1.0dB 2.0dB 3.0dB 4.0dB	The squelch function is set by bits 46 & 45. The center frequency gain of this element is 35dB; data-selected gain variations (-3.0dB to 4dB) are around this value.						
<u>36</u> 0 1	<u>Audio Output (RX)</u> - Output at bias, disabled - RX Audio Output Enabled	This function is designated for use in the RX mode, but can be used as a parallel output to the VCO Drive (TX).						
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	VCO Drive Amplifier - Gain Set -2.75dB -2.50dB -2.25dB -2.00dB -1.75dB -1.50dB -1.25dB -1.25dB -1.00dB -0.75dB -0.50dB 0.25dB 0.25dB 0.50dB 1.00dB 1.00dB	The in-line control amplifier/attenuator for the VCO TX channel drive output. This channel is also selected as Audio Output (RX) under the control of bit 36. This amplifier can be used in a volume control application.						
	Note: Bit 46 is transmitted first, bit 0 last.							

Circuit Descriptions and Serial Control Information

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Control bits	Element	Notes
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	VCO Drive Attenuator - Gain Set -44.8dB -44.8dB -41.6dB -38.4dB -35.3dB -32.0dB -28.8dB -25.6dB -22.4dB -19.2dB -16.0dB -12.8dB -9.6dB -6.4dB -3.2dB 0dB	The in-line control attenuator for the VCO TX channel drive output. This channel is also selected as the Audio Output (RX) under the control of bit 36. This attenuator can be used in a volume control application.
<u>27</u> 0	Output Control - Powersave VCO Ref. Amplifier ar Attenuator, disable Ref. and Drive Outputs. V _{BIAS} at both outputs. - Enable Ref. TX and VCO TX Outputs.	nd
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	VCO Reference Drive Amplifier - Gain Set -2.75dB -2.50dB -2.25dB -2.25dB -2.00dB -1.75dB -1.75dB -1.50dB -1.25dB -1.00dB -0.75dB -0.50dB -0.25dB 0dB 0.25dB 0.50dB 0.75dB 1.00dB 1.00dB	The in-line control amplifier/attenuator for the VCO reference channel drive output.
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	VCO Reference Drive Attenuator - Gain Set -28dB -24dB -20dB -16dB -12dB -12dB -8.0dB -4.0dB 0.0dB	The in-line control attenuator for the VCO reference channel drive output.
<u>19</u> 18 00 1 10 11	Output Drive Selector- Test signal path1- Process path2- Process path3- Process Bypass path4	To select the path to the output drive stages. Via de-emphasis element. Via de-emphasis bypass.
	<u>De-emphasis</u>	A selectable stage set around 1.0kHz with a characteristic of 6dB per octave.

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Control bits	Element	Notes
<u>17</u> 0 1	Post-Deviation Limiter Filter - Powersave and bypass filter - Enable filter path	
<u>16</u> 0 1	Post-Deviation Limiter Filter - Narrow filter bandwidth, cut-off = 2550Hz. - Wide filter bandwidth, cut-off = 3000Hz.	This selectable lowpass filter is adjust- able to Narrow (2550Hz) and Wide (3000Hz) bandwidths, allowing for different channel spacing requirements. See Figure 7.
	Deviation Limiter	A selectable, pre-set amplitude limiting stage for deviation control. Powersaved when Process Bypass is selected (bits 19 and 18).
15 0 1	<u>Pre-emphasis</u> - Powersave and bypass pre-emphasis. - Enable pre-emphasis path.	A selectable pre-emphasis stage set around 1.0kHz with a characteristic of 6dB per octave. It is powersaved when Process Bypass is selected (bits 19, 18).
	Post-process Gain	A fixed 2.0dB gain stage.
<u>14</u> 0 1	<u>Process Select</u> - Pre-process Internal Path - External Audio Process	To select either the Internal Signal path or the External Audio Process Input. This external input could be from a "Voice Scrambler" or the composite audio from a Sub-Audio or data signaling system.
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	Pre-process Gain - Gain Set -2.75dB -2.50dB -2.25dB -2.25dB -2.00dB -1.75dB -1.75dB -1.25dB -1.25dB -1.00dB -0.75dB -0.50dB 0.25dB 0.25dB 0.50dB 0.75dB 1.00dB	An in-line output drive stage that provides adjustable gain or attenuation to compensate for level tolerances in the external audio processes and peripherals. The output of this amplifier stage can be routed through an external path, or output to further voice processing (e.g. Frequency Inversion Voice Scrambling), or the introduction of Sub-Audible tones or In-Band data to the system.
9 0 1	Input Lowpass Filter - Bypass and powersave filter. - Enable Input Lowpass Filter.	A 3000Hz speech-path shaping low- pass filter. See Figure 5.
8 0 1	Input Highpass Filter - Bypass and powersave filter. - Enable Input Highpass Filter.	A 300Hz speech-path shaping highpass filter. See Figure 5. The filter gain is increased to +26dB when Mic. is selected (bits 2 and 1).
7 0 1	<u>Compressor (VOGAD)</u> - Bypass and powersave compressor. - Enable signal through compressor.	Pre-set, selectable voice amplitude compression circuitry that provides optimum drive levels to the transmission medium from differing signal level inputs. The Input Highpass Filter must be used with this element.

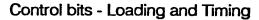
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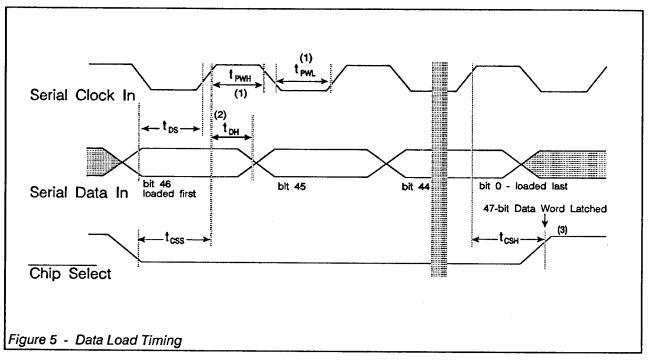
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Control bits	Element	Notes
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	Input Gain Amplifier - Gain Set 0.0dB 1.0dB 2.0dB 3.0dB 3.0dB 4.0dB 5.0dB 5.0dB 8.0dB 9.0dB 10.0dB 11.0dB 12.0dB 13.0dB 14.0dB 15.0dB 15.0dB	This is a gain element intended to adjust the drive level to the compressor, catering to differing signal sources and microphone sensitivities.
	Input Op-Amp	With external components, this amplifier can be used as a pre-amplifier and pre- emphasis circuit in the TX speech path. This gain element is available to adjust the drive level from differing signal sources and microphone sensitivities. Enabled by bits 2 and 1.
2010 01 10 11	Input Select - Internal bias path, no input selected. - Selected Microphone Input. Highpass Filter gain set to 26dB. - RX Audio Input. Powersave Op-Amp. - Internal bias path, no input selected. Powersave Input Op-Amp.	Transmit or receive audio sources are selected, inserting the appropriate path gain for the chosen input. The input path can be set to bias while allowing receiver monitoring. Input Op-Amp is enabled for transmit selection.
0 0 1	<u>Mic. Select</u> - Microphone Input 1 - Microphone Input 2	A Multiplexed "microphone" input allowing the use of differing type and level voice inputs.
	Note: Bit 46 is transmitted first,	, bit 0 last.
for use as a squeich detec	X Audio Input will also be available t tion level. Therefore the MX506 can l "significant noise" level is used to pr	hrough the squelch filter (when enabled) be set to "idle" with the majority of ele- roduce a "squelch drive output."
Test Signal Path This path, when selected, dynamically set and baland	can be used as a direct path, via the ce the VCO drive and reference outp	Output Drive Selector (bits 19 and 18), to put levels.
Pre-process Path: Post-process Path Attenuator and An	h: Pre-emphasis, Limiter, Post Deviati	ol: Highpass Filter, Input Lowpass Filter. ion Limiter Filter, De-emphasis, VCO Ref.
The following elements	s are active at all times:	

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The following elements are active at all times: Input Gain Amp, Pre-process Gain Amp, Post-process Amp (+2.0dB), VCO Drive Attenuator and Amplifier, Digital-to-Analog Converter, Squelch Level Comparator





Data Loading

Serial Data Bits, whose functions are described on Pages 6, 7, 8 and 9, are loaded to the MX506 using the timing format illustrated on this page. All 47 bits must be loaded. Data is loaded bit 46 first, bit 0 last.

Function Serial Clock (1)		Min.	Тур.	Max.	Unit
"High Pulse Width	t _{ewn}	600	-	-	ns
"Low" Pulse Width	t _{PWL}	600	-	-	ns
Serial Data (2)					
Data Set-up Time	t _{DS}	360	-	-	ns
Data Hold Time	t _{DH}	120	-	-	ns
Chip Select (3)					
Select Set-up Time	t _{css}	600	-	-	ns
Select Hold Time	t _{CSH}	600	-	- ,	ns

(1) The Serial Clock pulses do not have to be symmetrical, as shown above, but pulse lengths must conform to the "minimum" time specification.

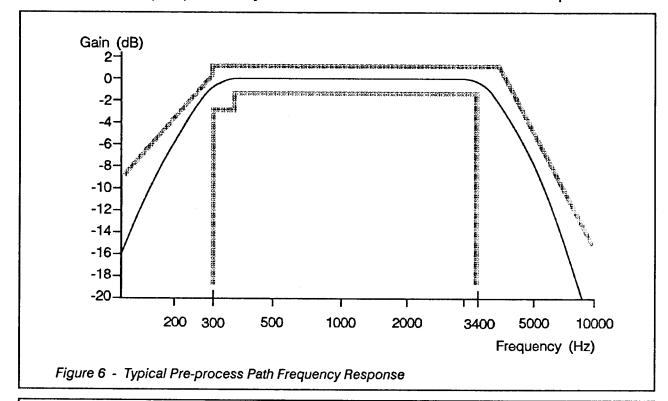
(2) Individual data bits (logic"1" or "0") are loaded to the device on the rising edge of the Input Serial Data clock pulse. The data hold period (t_{DH}) is to ensure that the data level is steady when it is sampled.

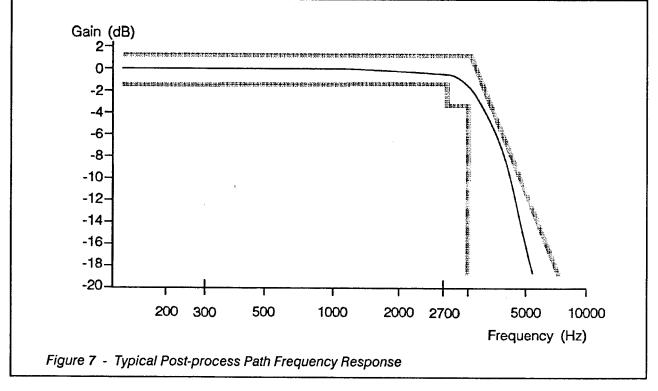
(3) The full 47-bit data word is latched into the device on the rising edge of the Chip Select waveform. At this time the loaded data is acted upon and the circuit configuration will change.

System Frequency Characteristics

Figure 6 shows a typical response curve of the Pre-process path (in receive mode) set against the device specification. The characteristic shape is produced by the Input Highpass and Lowpass Filters without the pre-emphasis element.

Figure 7 shows a typical response curve of the Post-process path set against the device specification. The characteristic shape is produced by the Post-deviation Limiter Filter without the de-emphasis element.





Specifications

Absolute Maximum Ratings

Exceeding the maximum rating can result in device damage. Operation of the device outside the operating limits is not implied.

Supply voltage	-0.3 to 7.0V
Input voltage at any pin	
(ref V _{ss} = 0V)	-0.3 to (V _{DD} + 0.3V) ±20mA
Sink/source current	±20mA
Total device dissipation	
@ 25°C	800mW Max.
Derating	10mW/°C
Operating temperature	-40°C to +85°C
Storage temperature	-55°C to 125°C

Operating Limits

All device characteristics are measured under the following conditions unless otherwise specified: ••••••

 $V_{DD} = 5.0V$

 $T_{AMB} = 25^{\circ}C$

Xtal/Clock $f_0 = 4.0 \text{ MHz}$

Audio level 0dB ref. = 600 mVrms (60% deviation, FM)

Characterist	ics Se	e Note	Min.	Тур.	Max.	Unit
Supply Voltage (۷)		4.5	5.0	5.5	V
Supply Current						•
All Elements Ena	bled		-	8.0	-	mA
Maximum Powers	save		-	3.0	-	mA
Dynamic Values						
Input Logic "1"		1	3.5	-	-	V
Input Logic "0"		1	-	-	1.5	V
Input Impedance						
Digital Input			0.1	1.0	-	MΩ
Mic 1 or 2			-	1.0	-	kΩ
RX Audio			50.0	-	-	kΩ
External Audio Pr	ocess		1.0	-	-	MΩ
Amp Input			1.0	-	-	MΩ
Noise, RSSI			1.0	-	-	MΩ
Output Impedance						
Pre-Process Audi	io		-	-	3.0	kΩ
Audio Out (RX)			-	-	3.0	kΩ
VCO Drive and R			-	-	3.0	έkΩ
Squelch Drive	Logic"1"		-	1.25	-	kΩ
	Logic"0"		•	150	-	Ω
Noise Output	Diode conducting		-	10.0	-	kΩ
	Diode not conducting		-	400	-	kΩ
_	h Isolation (Disabled)					
Switches			40.0	-	-	dB
Test Path			-	80.0	-	dB

Characteristics See Note Min. Typ. Max.

Unit

Signal Input Levels		11				
MIC 1 or 2			1.0	-	100	mVrms
RX Audio			-	123	200	mVrms
External Audio Process			-	-	795.0	mVrms
Noise, RSSI		2	-	-	5	ν
Signal Output Levels		11				
Pre-Process Audio			-	600	1000	mVrms
VCO (Drive, Ref.)			-	-	1000	mVrms
Audio (RX)			-	-	1000	mVrms
Variable Element Step To	lerance					
Input Gain amp			-	-	±30.0	%
Pre-Process Gain			-	-	±40.0	%
VCO Ref. Attenuator			-	-	±12.5	%
VCO Drive Attenuator			-	-	±15.625	%
VCO Amplifiers (Drive and	d Ref.)		-	-	±40.0	%
	•					
Output Distortion						
Output S/N		3	-	45.0	48.0	dB
THD		3	-	-	-40.0	dB
Compressor						
Dynamic Range			-	30.0	-	dB
Attack time			-	0.55	-	ms
Decay time			-	8.5	-	ms
Deviation Limiter						
Input thresholds		4	-	2.828	•	V p-р
						• •
Frequency Response						
Pre-Process Path (See Fig	gure 5)	6				
Passband Frequencies (-	3dB)		300	-	3400	Hz
Passband Ripple	(300-400Hz)	5	-3.0	-	1.0	dB
	(400-3400 Hz)	5	-1.5	- ,	1.0	dB
Stopband Attenuation	(f = 5 kHz)		3.0	4.2	-	dB
High Frequency Roll-off	(f >5kHz, <20kHz)		12.0	•	-	dB/oct
Stopband Attenuation	(f = 250 Hz)		-	2.3	-	dB
Low Frequency Roll-off	(f <250Hz)		6.0	-	•	dB/oct
		_				
Post-Process Path (See F		7				
Wideband: Lowpass Freq		_	-	-	3000	Hz
Passband Ripple	(<2700 Hz)	8	-1.5	-	1.0	dB
	(2700 -3000 Hz)	8	-3.0	-	1.0	dB

Characteristics	See	Note	Min.	Тур.	Max.	Unit
Stopband Attenuation	(f = 5 kHz)		12.2	17.0	-	dB
High Frequency Roll-off	(f >3 kHz, <20 kHz	z)	18	-	-	dB/oct
Narrowband: Lowpass Fr	equency (-3dB)		-	-	2550	Hz
Passband Ripple	(<2300 Hz)		-1.5	-	1.0	dB
	(2300-2550 Hz)		-3.0	-	1.0	dB
Stopband Attenuation	(f = 4.25 kHz)		12.2	17.0	-	dB
High Frequency Roll-off	(f >2300 Hz, <510	0 Hz)	18	-	-	dB/oct.
Pre-emphasis: Passband	Frequencies		300	-	3000	Hz
G	ain at 1kHz		-	0	-	dB
SI	ope Characteristic	9	-	6.0	-	dB/oct.
De-emphasis: Passband	Frequencies		300	•	3000	Hz
	ain at 1kHz		-	0	-	dB
SI	ope Characteristic	9	-	6.0	-	dB/oct.
Squelch Bandpass Filter						
Center Frequency Gain	(Wide and Narrow)		-	35.0	-	dB
Selectable Gain Narrow Band:	(8 x 1.0 dB steps)	10	-3.0	-	4.0	dB
Center Frequency	(fc)		-	18.75	-	kHz
Bandwidth	(fc±)		-	6.5	-	kHz
Wideband:						
Center Frequency	(fc)		-	25.5	-	kHz
Bandwidth	(fc±)		-	8.5	-	kHz

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Notes:

- 1. A percentage of the applied V_{DD} (70% or 30%).
- 2. These inputs are compared internally with the Digital-to-Analog converter.
- 3. With a signal output level of 600 mVrms, measured in a 30.0 kHz bandwidth.
- 4. Levels at the input of the Limiter element, centered around $\rm V_{\rm BIAS}$ (note 2).
- 5. This parameter remains within specification when pre-emphasis is employed.
- 6. With both input H.P.F. and L.P.F. in circuit, but without pre-emphasis.
- 7. With Limiter B.P.F., but without de-emphasis characteristics.
- 8. This parameter remains within specification when de-emphasis is employed.
- 9. Accuracy = ± 0.5 dB from nominal slope characteristic.
- 10. The gain variation around the center frequency (fc).
- 11. See Suggested Evaluation Tests for information on gain element settings.

Package Outline

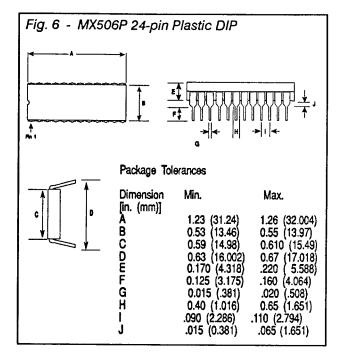
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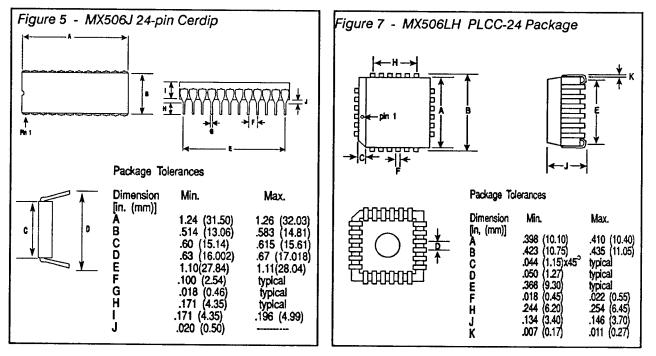
The MX506J Cerdip package is shown in Figure 5. The PDIP version is shown in Figure 6, and the LH in Figure 7.

For identification purposes the LH package has an ident spot adjacent to pin 1 and a chamfered corner between pins 3 and 4. Pins number counter-clockwise when viewed from the top (indent side).

Handling Precautions

The MX506LH is a CMOS LSI circuit which includes input protection. However, precautions should be taken to prevent static discharges which may cause damage.





Ordering Information

MX506J 24-pin Cerdip

MX506P 24-pin PDIP

MX506LH 24-lead Plastic Leaded Chip Carrier

Seoul, Korea ny odbunapbuoay Yoido P.O. Box 577, S-TEC International Co., Ltd. In Korea,

Fax: (02) 784-7500 Telex: K23456 STECI Phone: (02) 784-6800

Taipei, Taiwan, R.O.C. South Road Hyatt Tower 7F, No. 696 Tung Hua Corporation Mitronics International ,nswisT ni

Fax: (02) 776-1394 Telex: 12621 MITRONIC Phone: (02) 709-7626

COMMUNICATION MICROSYSTEMS

Whitham, Essex, England CM8 3TH 1 Wheaton Rd. (A member of CML Microsystems, Plc.) Consumer Microcircuits Ltd. in Europe,

Fax: (0376) 518247 Telex: 9932 CMICRO G Phone: (0376) 513833

Kawasaki 213, Japan 100-1, Sakado, Takatsu-ku R&D C-4F Kanagawa Science Park Teksel Co., Ltd. in Japan,

Fax: (044) 812-7433 Telex: 3842196 TEKSEL Phone: (044) 812-7430

Fax: (06) 391-8684 Phone: (06) 391-8661 Osaka office

in the Americas,

Winston-Salem, NC27105-1201 4800 Bethania Station Rd. WIT.I•XW

Intl. Telex: 5101012 SSSS 1202-117 (019) X67 1722-863 (008) Phone: (919) 744-5050



damaged by static discharge. MOS Device. May be **CAUTION**

Observe handling precautions



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Specifications subject to change.

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CML Product Data

In the process of creating a more global image, the three standard product semiconductor companies of CML Microsystems Plc (*Consumer Microcircuits Limited (UK), MX-COM, Inc (USA) and CML Microcircuits (Singapore) Pte Ltd)* have undergone name changes and, whilst maintaining their separate new names (*CML Microcircuits (UK) Ltd, CML Microcircuits (USA) Inc and CML Microcircuits (Singapore) Pte Ltd*), now operate under the single title CML Microcircuits.

These companies are all 100% owned operating companies of the CML Microsystems Plc Group and these changes are purely changes of name and do not change any underlying legal entities and hence will have no effect on any agreements or contacts currently in force.

CML Microcircuits Product Prefix Codes

Until the latter part of 1996, the differentiator between products manufactured and sold from MXCOM, Inc. and Consumer Microcircuits Limited were denoted by the prefixes MX and FX respectively. These products use the same silicon etc. and today still carry the same prefixes. In the latter part of 1996, both companies adopted the common prefix: CMX.

This notification is relevant product information to which it is attached.

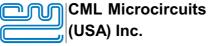
CML Microcircuits (USA) [formerly MX-COM, Inc.] Product Textual Marking

On CML Microcircuits (USA) products, the **'MX-COM'** textual logo is being replaced by a **'CML'** textual logo.

Company contact information is as below:



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